

# Cyber-Physical Systems

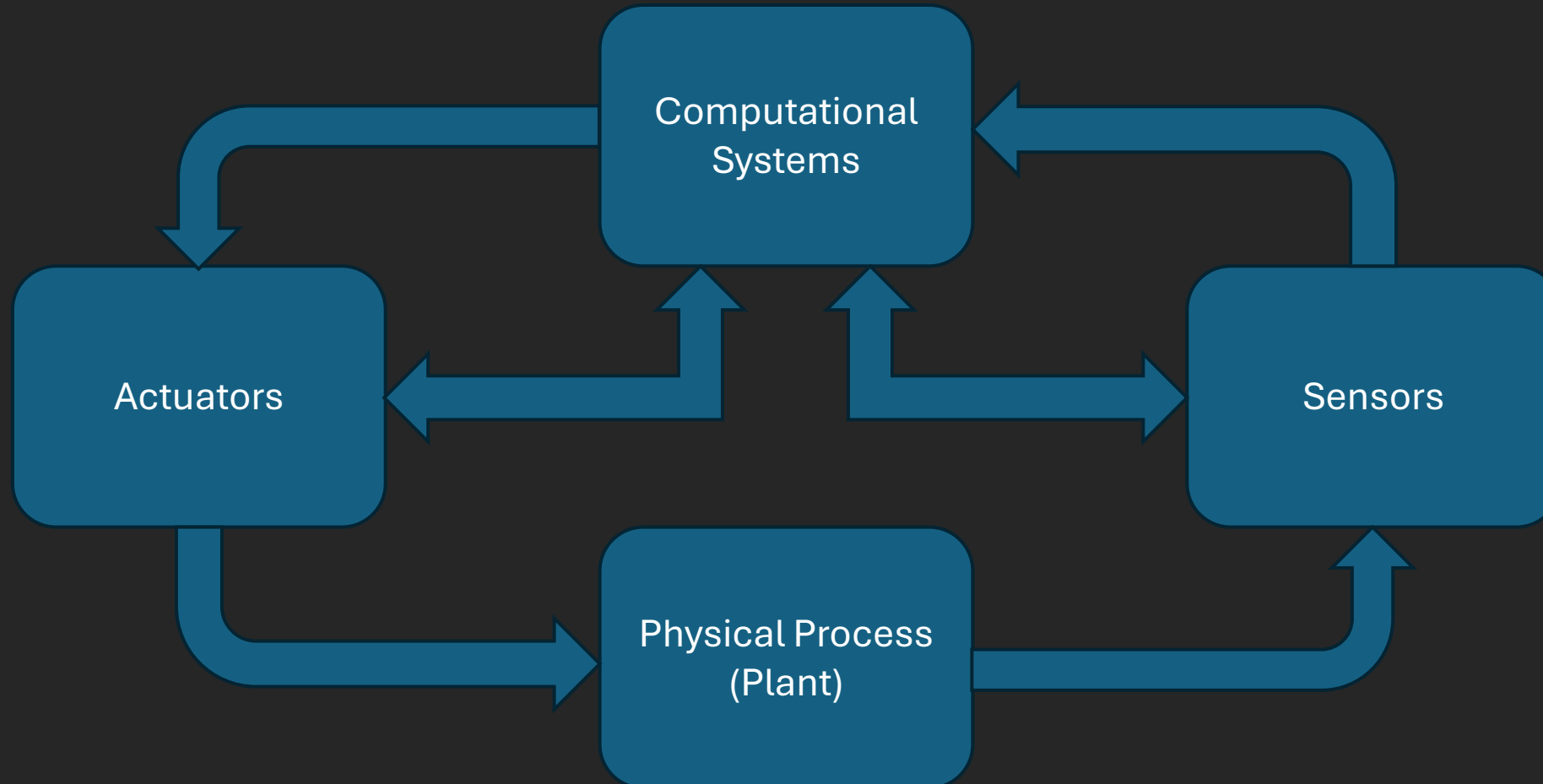
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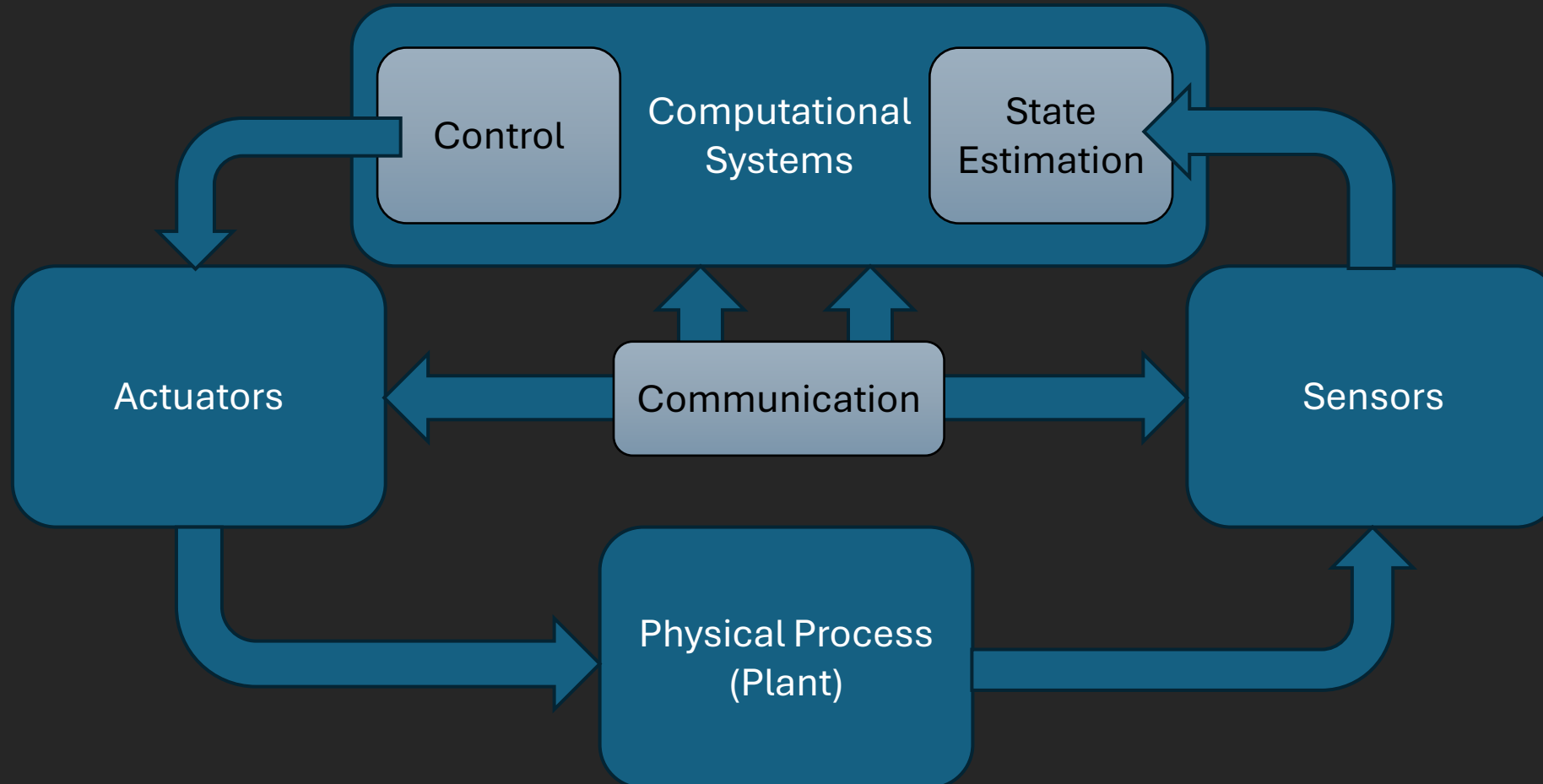
# Feedback Control



# What are Cyber-Physical Systems?



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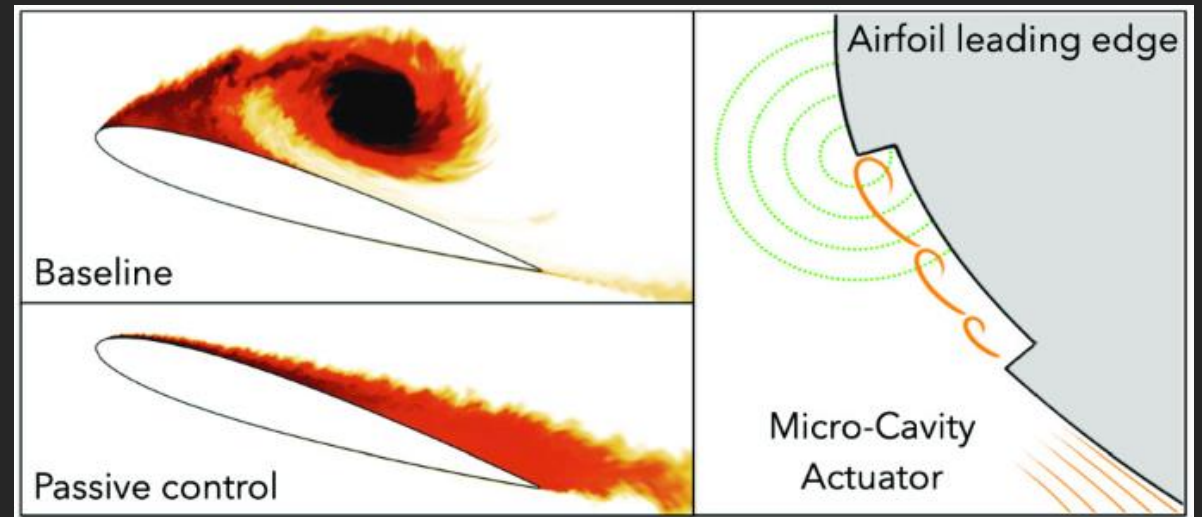


# Feedback Control

- Feedback control is a method of automatically regulating a system by continuously measuring its output, comparing it to a desired reference, and adjusting inputs to minimize the error
- Feedback control in Cyber-physical Systems
  - Computational components
  - System Models
  - State Estimation

# Non-Feedback Control

- Timers
- Rotational Stability – Spinning frisbee
- Passive drag elements (arrows and planes)
- Boiling food



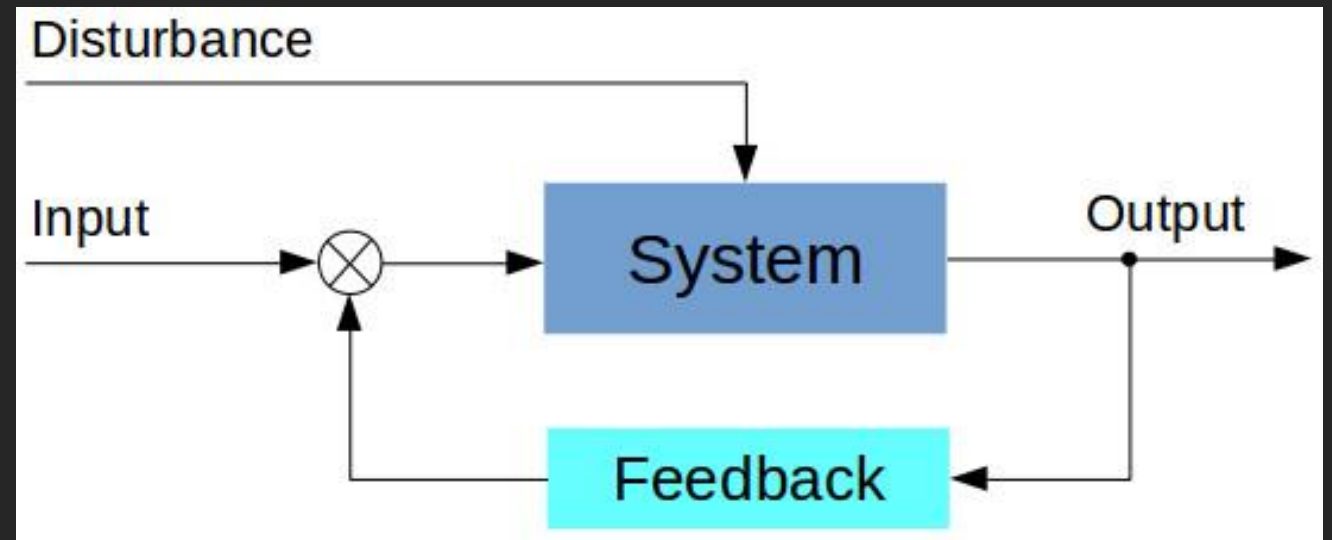
# Feedback Control

- Thermostat
- Cruise control
- Autofocus
- Computer fan speed



# Concepts of Feedback Control

- State-space control
- PID Control
- Other Control Concepts
  - Adaptive Control
  - Model Predictive Control
  - Distributed Control





# State Space Control



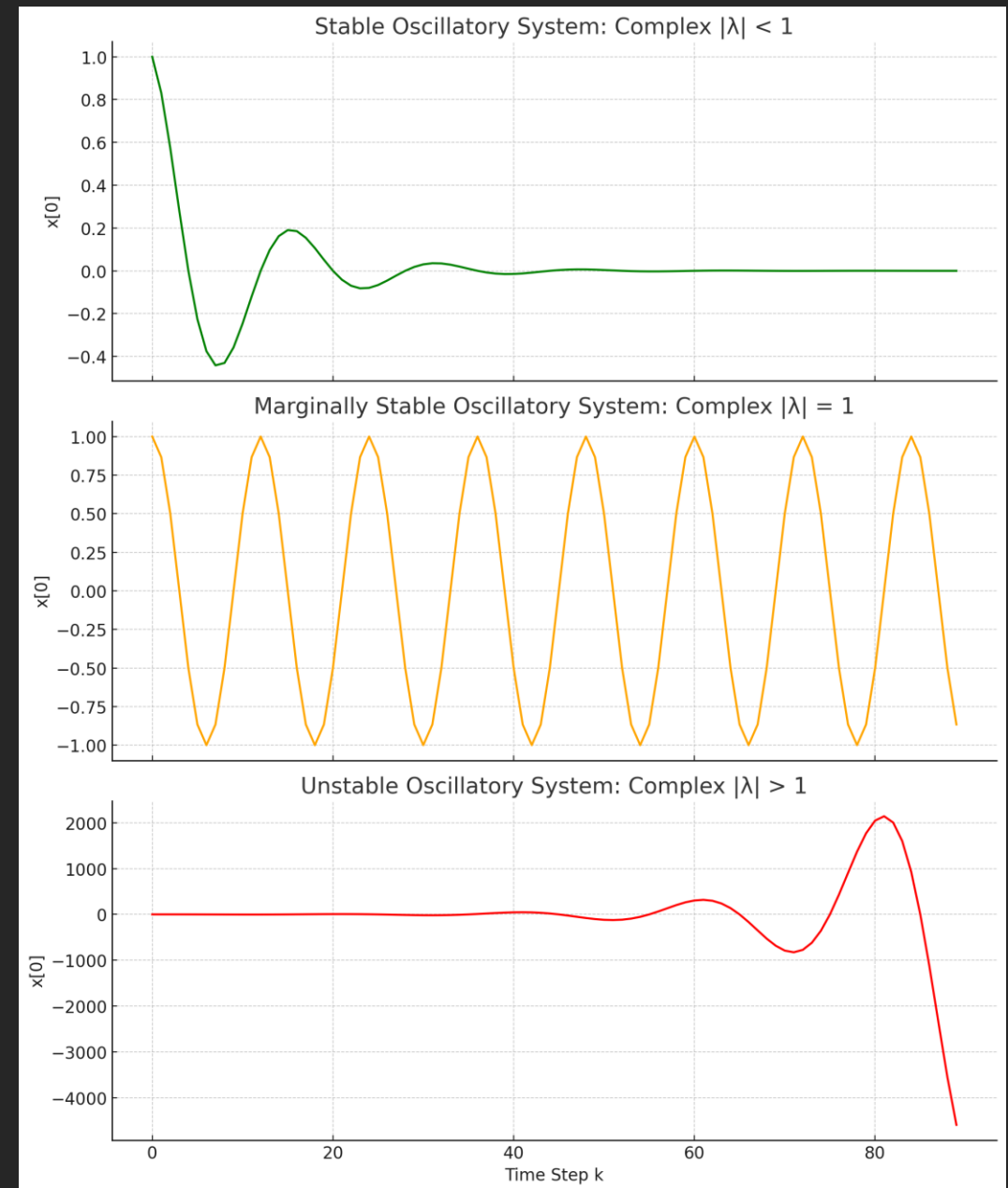
# State Space Control

- Canonical State Space Form (Assume full Observability)
  - $x_{k+1} = Ax_k + Bu_k$
  - $y_{k+1} = x_{k+1}$
- With no input system dynamics are governed by  $A$  for all time
  - $x_{k+1} = Ax_k$
- Stability analysis
  - Eigen values of  $A$  are given by  $\det(\lambda I - A) = 0$
  - Unstable if  $|\lambda_i| > 1$  for all  $\lambda_i$
  - Marginally Stable if at least one  $|\lambda_i| = 1$  and all other  $|\lambda_i| \leq 1$
  - Stable if all  $|\lambda_i| < 1$



# State Space Control

- $x_{k+1} = Ax_k$
- Stability analysis
  - $|\lambda_i| > 1$  for all  $\lambda_i$
  - At least one  $|\lambda_i| = 1$ , all other  $|\lambda_i| \leq 1$
  - $|\lambda_i| < 1$
- If stable, system will converge to dominate eigen value



# Can We Change the Eigan Values?

- The system dynamics can't change without altering the system.
- Create a control law
  - $x_{k+1} = Ax_k + Bu_k$
  - $u_k = -Kx_k$
- $x_{k+1} = Ax_k - BKx_k = (A - BK)x_k$
- Now system dynamics are governed by  $A - BK$  and its eigen values



# Control in State Space Methods

- Pole Placement
  - Choosing  $K$  such that the system has the exact dynamics you want
- Linear Quadratic Regulator
  - Chooses  $K$  to minimize a loss function
  - $J = \int_0^{\infty} x(t)^T Q x(t) + u(t)^T R u(t) dt$
  - $Q$  – penalizes deviation in the state
  - $R$  – penalizes control effort



# State Space Control

- Advantages
  - Tools for fine tuning exact system performance exist
  - Simple, predicable output
  - Works for multidimensional control
- Disadvantages
  - Requires exact knowledge of system
  - Errors in system dynamics could result in unstable outputs
  - Susceptible to noise
  - Not robust



# Model Predictive Control



# LQR State Space Control

- Linear Quadratic Regulator
  - Chooses  $K$  to minimize a loss function
  - $J = \int_0^{\infty} x(t)^T Q x(t) + u(t)^T R u(t) dt$
  - $Q$  – penalizes deviation in the state
  - $R$  – penalizes control effort
- Time horizon is infinite
- Once  $K$  is calculated, it is applied for all time



# Discrete Finite Time Horizon

- $x[k + 1] = f(x[k], u[k])$
- Look at a finite amount of time into the future
- $u^* = \arg \min_{u[0 \dots N-1]} \sum_{i=0}^{N-1} (x[i] - x_{ref}[i])^T Q (x[i] - x_{ref}[i]) + u[i]^T R u[i]$
- $u^* = \arg \min_{u[0 \dots N-1]} \sum_{i=0}^{N-1} x[i]_{error}^T Q x_{error} + u[i]^T R u[i]$

# Choosing $Q$ & $R$

$$\bullet Q = \begin{bmatrix} q_1 & 0 & \cdots & 0 \\ 0 & q_2 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \cdots & q_n \end{bmatrix}$$

$$\bullet R = \begin{bmatrix} r_1 & 0 & \cdots & 0 \\ 0 & r_2 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \cdots & r_n \end{bmatrix}$$

# Choosing $Q$ & $R$

- Usually,  $Q$  &  $R$  are diagonal because state variables are independent
- When to use non-diagonal matrices
  - Coupled state variables
  - Coupled control variables
- Example of coupled state variables
  - Angular speed and linear speed

$$\bullet \ x = \begin{bmatrix} x \\ y \\ \theta \\ v \\ \dot{\theta} \end{bmatrix}, Q = \begin{bmatrix} q_x & 0 & 0 & 0 & 0 \\ 0 & q_y & 0 & 0 & 0 \\ 0 & 0 & q_\theta & q_{v\theta} & 0 \\ 0 & 0 & q_{v\theta} & q_v & q_{v\dot{\theta}} \\ 0 & 0 & 0 & q_{v\dot{\theta}} & q_{\dot{\theta}} \end{bmatrix}$$

# Model Predictive Control

- Advantages
  - Handles multi-input, multi-output systems
  - Handles non-linear systems
  - Receding horizon approach and accurate system prediction
- Disadvantages
  - *Very* computationally expensive
  - Errors in system dynamics could result in unstable outputs
  - Tuning complexity
  - Implementation complexity

# Model Predictive Control



# PID Control



# Proportional Control

- Calculate error
  - $e(t) = x_{ref}(t) - x(t)$
- Calculate control value
  - $u(t) = K_p e(t)$
- What happens when there is a force resisting the system?
- Proportional control will always have steady state error

# Proportional-Integral Control

- Calculate error
  - $e(t) = x_{ref}(t) - x(t)$
- Calculate control value
  - $u(t) = K_P e(t) + K_I \int_0^t e(\tau) d\tau$
- Removes steady state error!
- Introduces overshoot.

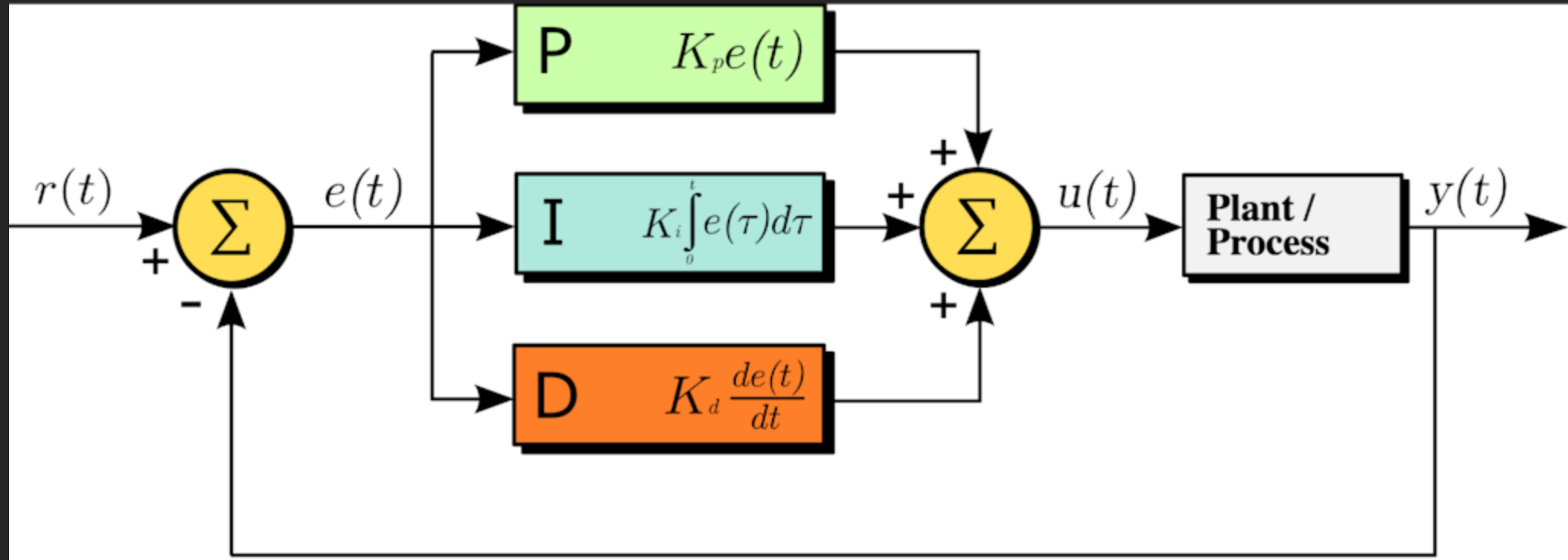


# Proportional-Integral-Derivative Control

- Calculate error
  - $e(t) = x_{ref}(t) - x(t)$
- Calculate control value
  - $u(t) = K_P e(t) + K_I \int_0^t e(\tau) d\tau + K_D \frac{d}{dt} e(t)$
- Can reduce overshoot
- Can produce instability

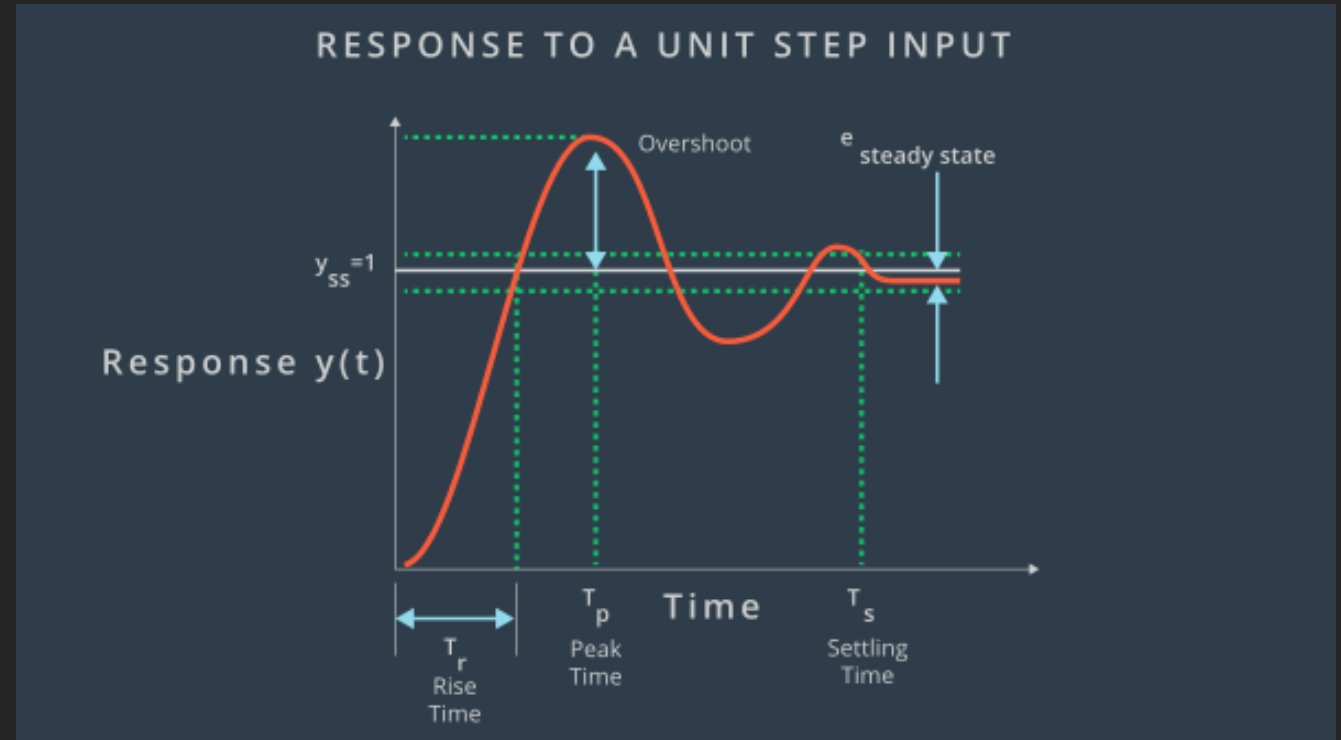
# PID Control in Discrete Time

- $u[k] = K_P + K_I \sum_{i=0}^k e[i]T_s + K_D \frac{e[k] - e[k-1]}{T_s}$



# Characterizing a PID Controller

- Rise Time
- Settling Time
- Overshoot
- Steady State Error

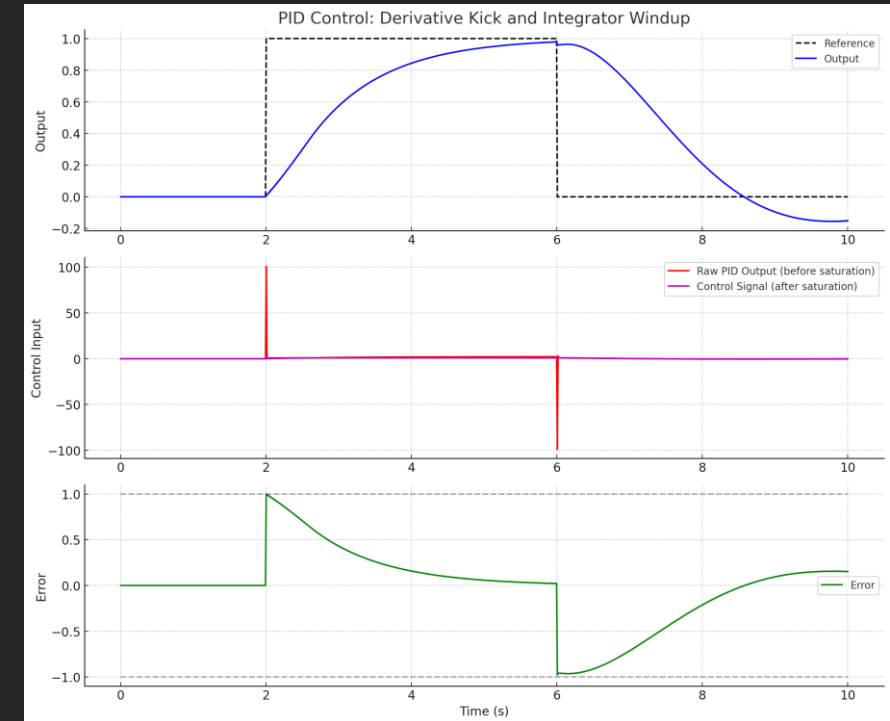


# PID Parameters

- $K_P$ 
  - Faster response, more reactive
  - Overshoot and oscillation
- $K_I$ 
  - Eliminates steady-state error
  - Sluggish, integrator windup
- $K_D$ 
  - Reduces overshoot, adds dampening
  - Sensitive to noise

# Potential Problems with PID

- Constant steady-state error
  - Integrator windup
- Derivative Kick
  - High frequency noise
  - Step change in reference signal
- Output Clamping
  - Calculate  $u(t)$  is greater than actuator can generate



# Derivative Kick

- $\frac{de}{dt} = \frac{d}{dt} (x_{ref}(t) - x(t))$
- $\frac{de}{dt} = \frac{dx_{ref}(t)}{dt} - \frac{dx(t)}{dt}$
- $\frac{dx_{ref}(t)}{dt}$  will spike if there is a change in reference signal
- $\frac{de}{dt} = - \frac{dx(t)}{dt}$

# Manually Tuning PID Controllers

- Start with  $K_I = 0$  and  $K_D = 0$
- Increase  $K_P$  until system begins to oscillate
- Add  $K_D$  to reduce overshoot and dampen oscillations
- Add  $K_I$  to eliminate steady-state error
- Repeat

# Ziegler-Nichols Method

- Set  $K_I = 0$  and  $K_D = 0$
- Increase  $K_P$  until system starts to oscillate
  - This gain is  $K_w$  and period of oscillation is  $T_w$

Controller	$K_P$	$K_I$	$K_D$
P	$0.5K_w$	-	-
PI	$0.45K_w$	$1.2K_P/T_w$	-
PID	$0.6K_w$	$2K_P/T_w$	$K_P T_w/8$



# PID Control

- Advantages
  - Simple – doesn't require detailed system modeling or design
  - Computationally efficient
  - Effective for many single input-single output systems
- Disadvantages
  - Limited in multi-variable systems
  - Poor predictive capability
  - Sensitive to tuning and nonlinearities

# Adaptive Control



# Adaptive Control

- Challenges in implementing feedback control systems
  - Unknown system dynamics
  - Changing system dynamics
- Examples
  - Vehicles carry different payloads
  - Electrical grid demand changes from season to season
  - User of a system might change

# Adaptive Control

- $x_{k+1} = A_k x_k + B_k u_k$
- Adaptive control
  - As you control your system, save the last N data point
  - Perform linear regression to estimate  $A_k$  and  $B_k$
  - Use  $A_k$  and  $B_k$  to calculate new control law

# Adaptive control

- Self Tuning Regulator
  - Estimates  $A_k$  and  $B_k$  in real time
  - Recalculates control law (pole-place, MPC, PID) based on new  $A_k$  and  $B_k$
- Gain Scheduling (Quasi-Adaptive)
  - Precomputes different laws (pole-place, MPC, PID)
  - Switches or interpolates between them based on estimated  $A_k$  and  $B_k$

# Adaptive Control



# Adaptive Control

